

BUS INTERFACE FOR TRANSFER OF SONET/SDH DATA

FIELD

This invention relates generally to a bus interface used
5 to transfer SONET/SDH data. More specifically, this
invention relates to a bus interface for the transfer of
SONET/SDH data over a serial backplane.

BACKGROUND OF THE INVENTION

10 In conventional data communication systems, a local node
with a large number of SONET/SDH terminations, may process a
large variety of payload types. Common payloads are
Asynchronous Transfer Mode (ATM), Packet Over SONET (POS),
and Time Division Multiplexing (TDM) traffic. In general,
15 each of the payload types is processed by specialised
hardware residing in disparate cards. For traffic not
terminating in the local node, the traffic may be groomed and
transported from input fibers to arbitrary output fibers.

20 The Combus standard provides a common interface between
SONET termination devices and payload processing devices.
However, the Combus standard is limited to OC-3 streams and
contains 11 signals per interface. For a high capacity node,
the number of signals required typically exceeds the limits
25 of the Combus standard.

Another existing approach to connecting SONET/SDH termination devices to payload processing devices requires the reconstruction of a serial SONET/SDH stream post SONET/SDH termination. However, this method suffers from the disadvantage of requiring duplicate SONET/SDH processing at the payload processing devices. Another disadvantage of this method is the lack of guaranteed transitions on the serial links as SONET/SDH scrambling only provides a statistical amount of transitions. As a result, complex clock and data recovery phase-locked loops are often required.

To overcome the lack of guaranteed transitions on the serial links, an alternative approach is used that involves reconstructing a serial SONET/SDH stream post SONET/SDH termination and then applying a line code that guarantees transitions on the serial link. Again, this method suffers from the disadvantage of requiring duplicate SONET/SDH processing at the payload processors.

It is, therefore, an object of this invention to provide an improved bus interface to connect SONET/SDH termination devices with payload processing devices.

It is a further object of this invention to provide a bus interface to connect SONET/SDH termination devices with

payload processing devices utilizing a minimum number of signals.

It is still a further object of this invention to
5 provide a bus interface that can scale with future advances
in bandwidth in serial link technology.

SUMMARY OF THE INVENTION

These and other objects of the invention are provided in
10 a new and improved method of connecting SONET/SDH termination
devices with payload processing devices while requiring a
minimum number of signals. The protocol used in the method
allows the SONET/SDH termination device to handoff at the
SONET line termination level, path termination level, and
15 tributary termination level (SDH multiplex section
termination level, high-order path terminating level and low-
order path terminating level). The protocol is also capable
of scaling with future advances in bandwidth in serial link
technology.

20 The method consists of providing a transmit interface
and a receive interface. In operation, the transmit
interface receives an incoming SONET/SDH signal stream and
converts the SONET/SDH signal stream into outgoing low
25 voltage differential signal (LVDS) levels. The SONET/SDH
signal streams are mapped into 8B/10B control characters to

label the SONET/SDH frame boundaries. Potential SONET/SDH frame boundaries include transport frame, high-order path frame and low-order path frame boundaries.

5 The receive interface receives incoming LVDS signal levels and converts the LVDS levels into and outgoing SONET/SDH signal streams. The 8B/10B control characters labeling the SONET/SDH frame boundaries are decoded into SONET/SDH control signals.

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By mapping a descrambled SONET/SDH data stream into 8B/10B control characters, proper data transitions on serial links can be ensured. Also, the mapping preserves the DC balance.

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Preferably, the 8B/10B control characters that have an even number of ones and zeros have their positive and negative disparity codes treated as separate control characters. Line code violations of these 8B/10B characters may be used to monitor error performance of serial links.

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Alternatively, the signals may be stored in a buffer. The signals can then be transferred using a universal frame pulse with a software programmable delay to allow the transfer of a single SONET/SDH signal over multiple links.

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Preferably, the method also includes providing transparent in-band error reporting such that errors detected at a SONET/SDH receiver can be transferred to a transmitter to construct remote error and defect indication codes. The
5 method may also include inserting a pseudo-random bit sequence pattern in serial transmit links to allow data path verification prior to injection of an actual payload.

Alternatively, the method may include overwriting one of
10 the E1 and B1 bytes to form a pattern. This pattern allows in-service monitoring of link functionality as well as monitoring of downstream cross-connect mis-configurations. Optionally, the bytes in E1 may be overwritten with the complement of a value in B1 bytes.

15 The invention also includes a bus interface device operative to perform the steps of the method described above.

Other objects and advantages of the invention will
20 become clear from the following detailed description of the preferred embodiment, which is presented by way of illustration only and without limiting the scope of the invention to the details thereof.

BRIEF DESCRIPTION OF THE DRAWINGS

Many objects and advantages of the present invention will be apparent to those of ordinary skill in the art when this specification is read in conjunction with the attached
5 drawings wherein like reference numerals are applied to like elements and wherein:

Fig. 1 is a schematic block diagram depicting an embodiment of the bus interface;

10 **Fig. 1A** is a table containing 8B/10B control characters for a transmit interface;

15 **Fig. 1B** is a table containing 8B/10B control characters for a transmit interface;

Fig. 2 is a block diagram depicting an embodiment of an incoming data 8B/10B encoder block;

20 **Fig. 3** is a block diagram depicting an embodiment of a transmit time-slot interchange block;

Fig. 4 is a block diagram depicting an embodiment of a receive 8B/10B TeleCombus decoder block;

Fig. 5A is a table containing 8B/10B control characters for a receive interface;

Fig. 5B is a table containing 8B/10B control characters
5 for a receive interface;

DETAILED DESCRIPTION

Referring to **Fig. 1**, the bus interface **10** is depicted having a transmit interface **20** and a receive interface **22**.

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The transmit interface **20** is comprised of an Incoming TeleCombus PRBS Processor (ITPP) block **30**, an Incoming Data 8B/10B Encoder (ID8E) block **34**, an Incoming PRBS 8B/10B Encoder (IP8E) block **32**, Transmit Time-Slot Working, Protect and Auxiliary Interchange (TTSI) blocks **36**, **38** and **40**,
15 respectively, Transmit Working, Protect and Auxiliary Disparity Encoder (TRDE) blocks **42**, **44** and **46**, respectively, Transmit Working, Protect and Auxiliary Serialiser (PISO) blocks **48**, **50** and **52**, respectively, and Transmit Working,
20 Protect and Auxiliary LVDS Interface blocks **54**, **56** and **58**, respectively. The acronym PRBS refers to a pseudo-random bit sequence.

Incoming TeleCombus PRBS Processor

The Incoming TeleCombus PRBS Processor (ITPP) block **30**
25 provides in-service and off-line diagnostics of the incoming

TeleCombus stream and equipment downstream of the three sets of transmit LVDS links. Within the ITPP block **30**, a total of four ITPP sub-blocks (ITPP #1-#4) are instantiated in the Telecom Bus Serialiser (TBS) device. Each ITPP sub-block has the capacity to monitor and source Pseudo Random Bit Sequence (PRBS) data of an STS-12/STM-4 stream. A set of four ITPP blocks may be connected in tandem to service an STS-48c/STM-16-16c stream.

The Incoming TeleCombus PRBS Processor (ITPP) block **30** includes a PRBS detector and a PRBS generator.

PRBS Detector

Each ITPP sub-block has an independent PRBS detector and generator. The PRBS detector in ITPP #1 to ITPP #4 monitors the four sections of the incoming data stream ID[1][7:0] to ID[4][7:0], respectively. When enabled, the PRBS detector monitors synchronous payload envelope (SPE) / higher order virtual container (VC3 or VC4-Xc) bytes in the incoming data stream. The incoming data is compared against the expected value derived from an internal linear feedback shift register (LFSR) with a polynomial of $X^{23} + X^{18} + 1$. If the incoming data fails to match the expected value for three consecutive bytes, the PRBS detector will enter out-of-synchronization (OOS) state. The LFSR will be re-initialized using the incoming data bytes. The new LFSR seed is confirmed by

comparison with subsequent incoming data bytes. The PRBS detector will exit the OOS state when the incoming data matches the LFSR output for three consecutive bytes. The PRBS detector will remain in the OOS state and re-load the LFSR if confirmation failed. The PRBS detector counts PRBS byte errors and optionally generates interrupts when it enters and exits the OOS state.

The PRBS detector may be configured to also monitor the B1 and E1 bytes in the incoming data stream. The B1 byte in each incoming STS-1/STM-0 is compared with an independently software programmable value. The E1 byte is compared with the complement of the programmable value. An interrupt is optionally generated when there is a change from the matched to mismatched state and vice-versa. The incoming B1 bytes are captured in a set of software readable registers. This facility allows in-service diagnosis of provisioning errors in upstream cross-connect devices.

PRBS Generator

The PRBS generator in ITPP #1 to ITPP #4 may optionally overwrite the data in incoming data stream ID[1][7:0] to ID[4][7:0], respectively. When enabled, the PRBS generator inserts synchronous payload envelope (SPE) / higher order virtual container (VC3 or VC4-Xc) bytes into the serial transmit links. The inserted data is derived from an

internal linear feedback shift register (LFSR) with a polynomial of $X^{23} + X^{18} + 1$.

The PRBS generator may be configured to optionally insert a software programmable byte into the B1 byte of each STS-1/STM-0 stream the serial transmit links. The E1 bytes may be over-written to the complement of the value inserted into the B1 bytes. This facility allows in-service diagnosis of provisioning errors in downstream cross-connect devices.

Incoming Data 8B/10B Encoder

The Incoming Data 8B/10B Encoder (ID8E) block **34** constructs an 8B/10B character stream from an incoming TeleCombus carrying an STS-12/STM-4 stream. Within the ID8E block **34**, a total of four ID8E sub-blocks (ID8E #1 to #4) are instantiated in the TBS device. ID8E sub-blocks #1 to #4 process incoming data streams ID[1][7:0] to ID[4][7:0], respectively.

Frame Counter

Each of the ID8E sub-blocks #1 to #4 of ID8E block **34** includes a Frame Counter **102** and an 8B/10B Encoder **104** as shown in **Figure 2**.

The Frame Counter **102** keeps track of the octet identity of the incoming data stream. The Frame Counter **102** is initialized by the J0 pulse on the IJ0J1 and IPL signals. It identifies the positive stuff opportunity (PSO) and negative stuff opportunity (H3) bytes within the transport frame so that high-order path pointer justification events can be identified and encoded.

8B/10B Encoder

The 8B/10B Encoder **104** converts bytes in the incoming STS-12/STM-4 stream to 8B/10B characters. It can operate in one of three modes; multiplex section termination (MST), high-order path termination (HPT) and low-order path termination (LPT) modes. The modes relate to the level of SONET/SDH processing capability in the external device driving the incoming TeleCombus (ID[4:1][7:0]).

In MST mode, the upstream device is a multiplex section terminator. It has identified transport frame boundaries.

The first J0 byte (J0) is encoded by an 8B/10B control character. Incoming TeleCombus signals ITV5[4:1], ITPL[4:1], and ITAIS[4:1] and the J1 portion of IJ0J1[4:1] are ignored.

In HPT mode, the upstream device is a high-order path terminator and has performed pointer processing to identify STS/AU level pointer justification events. It has processed

all the STS/VC3/VC4 path overhead bytes. The H3 bytes in the absence of negative pointer justification events, the PSO byte in the presence of positive pointer justification events may be encoded. Alternately, the J1 byte may be encoded.

- 5 Incoming TeleCombus signals ITV5[4:1], ITPL[4:1], and ITAIS[4:1] are ignored.

In LPT mode, the upstream device is a low-order path terminator. It has performed tributary level pointer processing to identify tributary payload bytes and have terminated tributary payload / low-order virtual container bytes. In addition to MST and HPT mode bytes, V5 bytes and all bytes that are not part of a low-order path payload are encoded. Note that in drop-and-continue operation, the TBS must be configured to regard the upstream device as one appropriate for the continued path.

Figures 1A and **1B** show the mapping of TeleCombus control bytes and signals into 8B/10B control characters. **Figure 1A** shows the character mapping for the MST and HPT modes of operation. **Figure 1B** shows the character mapping for the LPT mode of operation.

Incoming PRBS 8B/10B Encoder

The Incoming PRBS 8B/10B Encoder (IP8E) block **32** constructs an 8B/10B character stream from the output of the

ITPP block **30**. Within the IP8E block **32**, a total of four IP8E sub-blocks (IP8E #1 to #4) are instantiated in the TBS device. IP8E sub-blocks #1 to #4 process data from ITPP sub-blocks #1 to #4, respectively. The IP8E block **32** may be
5 functionally identical to the ID8E block **34**.

Transmit Time-slot Interchange

The Transmit Time-slot Interchange (TTSI) blocks **36**, **38** and **40** re-arrange the constituent STS-1/STM-0 streams of an
10 STS-48/STM-16 stream in a software configurable order. The TTSI blocks **36**, **38** and **40** also support multi-casting where an incoming STS-1/STM-0 stream is placed on two or more outgoing time-slots. The Transmit Working Time-slot Interchange (TWTI) block **36** performs time-slot re-arrangement for data
15 destined for the working transmit LVDS links (TPWRK[4:1]/TNWRK[4:1]). The Transmit Protection Time-slot Interchange (TPTI) block **38** services the protection transmit LVDS links (TPPROT[4:1]/TNPROT[4:1]) while the Transmit Auxiliary Time-slot Interchange (TATI) block **40** services the
20 auxiliary transmit LVDS links (TPAUX[4:1]/TNAUX[4:1]).

Each of the TTSI blocks **36**, **38** and **40** includes a Data Buffer **106** and Connection Memory **108**.

Data Buffer

The Data Buffer **106** contains a double buffer structure. The incoming data stream is first loaded into an input shift register. A frame counter initiates a transfer of the data
5 to the holding register once all 48 constituent STS-1/STM-0 streams have been shifted in. The data is read out of the holding register in the order specified by the Connection Memory **108**.

10 Connection Memory

The Connection Memory **108** contains two mapping pages: page 0 and page 1. One page is designated the active page and the other the stand-by page. Selection between which page is to be active and which is to be stand-by is
15 controlled by the TCMP signal. The Connection Memory **108** samples the value on the TCMP signal at the J0 byte position of the incoming data stream and swaps the active/standby status of the two pages at the first A1 byte of the next frame. This arrangement allows all devices in a cross-
20 connect system to be updated in a coordinated fashion. Consequently, STS-1/STM-0 streams not being assigned new time-slots are unaffected by page swaps.

Transmit 8B/10B Running Disparity Encoder

25 The Transmit 8B/10B Running Disparity Encoder (TRDE) blocks **42**, **44** and **46** correct the running disparity of an

8B/10B character stream. The input data to the TRDE blocks **42**, **44** and **46** originates from either the ID8E sub-blocks #1 to #4 or the IP8E sub-blocks #1 to #4 at which point they have correct running disparity. However, due to the time-
5 slot re-arrangement activities of the TTSI blocks **36**, **38** and **40**, the running disparity is no longer consistent. The TRDE block inverts the 6B and 4B sub-characters to ensure correct running disparity.

10 There are a total of twelve TRDE sub-blocks instantiated in the TBS device. Four TRDE sub-blocks (TWDE #1 to #4), within Transmit Working Disparity Encoder **42** are dedicated to the working transmit LVDS links (TPWRK[4:1]/TNWRK[4:1]). The Transmit Protection Disparity Encoder **44** (TPDE #1 to #4)
15 corrects running disparity for characters destined for the protection transmit LVDS links (TPPROT[4:1]/TNPROT[4:1]) while the Transmit Auxiliary Disparity Encoder **46** (TADE #1 to #4) services the auxiliary transmit LVDS links (TPAUX[4:1]/TNAUX[4:1]).

20 Transmit Serialiser

The Transmit Serialiser (PISO) blocks **48**, **50** and **52** convert 8B/10B characters to bit-serial format. There are a total of twelve PISO sub-blocks instantiated in the TBS
25 device. Four PISO sub-blocks, Transmit Working Serialiser **48** (TWPS #1 to #4) are dedicated to the working transmit LVDS

links (TPWRK[4:1]/TNWRK[4:1]). The Transmit Protection
Serialiser **50** (TPPS #1 to #4) generates serial streams for
the protection transmit LVDS links (TPPROT[4:1]/TNPROT[4:1])
while the Transmit Auxiliary Serialiser **52** (TAPS #1 to #4)
5 are associated with the auxiliary transmit LVDS links
(TPAUX[4:1]/TNAUX[4:1]).

LVDS Transmitter

The LVDS Transmitters, TWLV block **54**, TPLV block **56** and
10 TALV block **58** (referred to as TXLV blocks) convert 8B/10B
encoded digital bit-serial streams to LVDS signaling levels.
A total of twelve TXLV sub-blocks are instantiated in the TBS
device. Four TXLV sub-blocks, Transmit Working LVDS
Interface **54** (TWLV #1 to #4) drives the working transmit LVDS
15 links (TPWRK[4:1]/TNWRK[4:1]). The Transmit Protection LVDS
Interface **56** (TPLV #1 to #4) drives the protection transmit
LVDS links (TPPROT[4:1]/TNPROT[4:1]) while the Transmit
Auxiliary LVDS Interface **58** (TALV #1 to #4) are associated
with the auxiliary transmit LVDS links
20 (TPAUX[4:1]/TNAUX[4:1]).

Clock Synthesis Unit

The Clock Synthesis Unit (CSU) block **62** generates the
777.6 MHz clock for the transmit and receive LVDS links.

25

Transmit Reference Generator

The Transmit Voltage Reference Generator block **60** generates bias voltages and currents for the LVDS Transmitters.

5

Receive Interface

The receive interface **22** is comprised of Receive LVDS Interface blocks **64**, **66**, and **68**, Data Recovery Units **70**, **72** and **74**, Receive 8B/10B Decoders **76**, **78** and **80**, Receive PRBS Monitors **82**, **84** and **86**, Receive Time-Slot Interchanges **88**, **90** and **92**, and an Outgoing TeleCombus PRBS Generator **94**.

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LVDS Receivers

The LVDS Receivers, RWLV block **64**, RPLV block **66** and RALV block **68** (referred to as RXLV blocks) convert LVDS signaling levels to 8B/10B encoded digital bit-serial. A total of twelve RXLV sub-blocks are instantiated in the TBS device. Four RXLV sub-blocks, Receive Working LVDS Interface **64** (RWLV #1 to #4) connect to the working receive LVDS links (RPWRK[4:1]/RNWRK[4:1]). The Receive Protection LVDS Interface **66** (RPLV #1 to #4) connects to the protection receive LVDS links (RPPROT[4:1]/RNPROT[4:1]) while the Receive Auxiliary LVDS Interface **68** (RALV #1 to #4) are associated with the auxiliary receive LVDS links (RPAUX[4:1]/RNAUX[4:1]).

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Data Recovery Units

The Data Recovery Unit (DRU) blocks **70**, **72** and **74** monitor the receive LVDS link for transitions to determine the extent of bit cycles on the link. They then adjust its internal timing to sample the link in the middle of the data "eye". A total of twelve DRU sub-blocks are instantiated in the TBS device. Four DRU sub-blocks, Working Data Recovery Units **70** (WDRU #1 to #4) retrieves data from the working receive LVDS links (RPWRK[4:1]/RNWRK[4:1]). The Protection Data Recovery Units **72** (PDRU #1 to #4) process the protection receive LVDS links (RPPROT[4:1]/RNPROT[4:1]) while the Auxiliary Data Recovery Units **74** (RALV #1 to #4) are associated with the auxiliary receive LVDS links (RPAUX[4:1]/RNAUX[4:1]).

The DRU blocks also convert the bit serial stream into 10-bit words. The words are constructed from ten consecutive received bits without regard to 8B/10B character boundaries.

Receive 8B/10B TeleCombus Decoder

The Receive 8B/10B TeleCombus Decoder (R8TD) blocks **76**, **78** and **80** frame to the receive stream to find 8B/10B character boundaries. They also contain a FIFO to bridge between the timing domain of the receive LVDS links and the system clock timing domain. A total of twelve R8TD sub-

blocks are instantiated in the TBS device. Four R8TD sub-blocks, Receiver Working 8B/10B Decoder blocks **76** (RW8D #1 to #4) perform framing and elastic store functions on data retrieved from the working receive LVDS links

5 (RPWRK[4:1]/RNWRK[4:1]). The Receive 8B/10B Decoder blocks **78** (RP8D #1 to #4) process data on the protection receive LVDS links (RPPROT[4:1]/RNPROT[4:1]) while the Receive Auxiliary 8B/10B Decoder blocks **80** (RA8D #1 to #4) are associated with the auxiliary receive LVDS links
10 (RPAUX[4:1]/RNAUX[4:1]).

Each of the R8TD sub-blocks includes a FIFO buffer **110**, a Frame Counter **112**, a Character Aligner **114**, a Frame Aligner **116** and a Character Decoder **118** shown in **Figure 4**.

FIFO Buffer

The FIFO buffer **110** provides isolation between the timing domain of the associated receive LVDS link and that of the system clock (SYSCLK). Data with arbitrary alignment to
20 8B/10B characters are written into a 10-bit by 24-word deep FIFO buffer **110** at the link clock rate. Data is read from the FIFO buffer **110** at every SYSCLK cycle.

Frame Counter

25 The Frame Counter **112** keeps track of the octet identity of the outgoing data stream. It is initialized by a delayed

version of the RJ0FP signal. It identifies the positive stuff opportunity (PSO) and negative stuff opportunity (H3) bytes within the transport frame so that high-order path pointer justification events can be identified and decoded.

5

Character Aligner

The Character Aligner **114** locates character boundaries in the incoming 8B/10B data stream. The framer logic may be in one of two states, SYNC state and HUNT state. It uses the 8B/10B control character (K28.5) used to encode the SONET/SDH J0 byte to locate character boundaries and to enter the SYNC state. It monitors the receive data stream for line code violations (LCV). An LCV is declared when the running disparity of the receive data is not consistent with the previous character or the data is not one of the characters defined in IEEE std. 802.3. Excessive LCVs are used to transition the framer logic to the HUNT state.

Normal operation occurs when the character aligner **114** is in the SYNC state. 8B/10B characters are extracted from the FIFO buffer **110** using the character alignment of the K28.5 character that caused entry to the SYNC state. Mimic K28.5 characters at other alignments are ignored. The receive data is constantly monitored for line code violations. If 5 or more LCVs are detected in a window of 15 characters, the character aligner transitions to the HUNT

state. It will search all possible alignments in the receive data for the K28.5 character. In the mean time, the original character alignment is maintained until a K28.5 character is found. At that point, the character alignment is moved to this new location and the character aligner transitions to the SYNC state.

Frame Aligner

The frame aligner **116** monitors the data read from the FIFO buffer **110** for the J0 byte. When the frame counter **112** indicates the J0 byte position, a J0 character is expected to be read from the FIFO buffer **110**. If a J0 byte is read out of the FIFO buffer **110** at other byte positions, a J0 byte error counter is incremented. When the counter reaches a count of 3, the frame aligner **116** transitions to HUNT state. The next time a J0 character is read from the FIFO buffer **110**, the associated read address is latched and the frame aligner **116** transitions back to the SYNC state. The J0 byte error counter is cleared when a J0 byte is read from the FIFO buffer **110** at the expected position.

Character Decoder

The character decoder **118** decodes the incoming 8B/10B control characters into an extended set of TeleCombus control signals. **Figures 5A** and **5B** show the mapping of 8B/10B

control characters into TeleCombus control signals. **Figure 5A** shows the character mapping for the MST and HPT modes of operation and **Figure 5B** shows the character mapping for the LPT mode of operation in the 8B/10B encoder in an external device upstream of the TBS. The character decoder itself is not mode sensitive.

Receive PRBS Monitor

The Receive PRBS Monitor (RPRM) blocks **82**, **84** and **86** provide in-service and off-line diagnostics of the receive LVDS links. A total of twelve RPRM sub-blocks are instantiated in the TBS device. Four RPRM sub-blocks, Receive Working PRBS Monitor **82** (RWPM #1 to #4) connect to the working receive LVDS links (RPWRK[4:1]/RNWRK[4:1]). The Receive Protection PRBS Monitor **84** (RPPM #1 to #4) connect to the protection receive LVDS links (RPPROT[4:1]/RNPROT[4:1]) while the Receive Auxiliary PRBS Monitor **86** (RAPM #1 to #4) are associated with the auxiliary receive LVDS links (RPAUX[4:1]/RNAUX[4:1]). The RPRM blocks **82**, **84** and **86** are functionally identical to the monitor section of the ITPP block **30**.

Receive Time-slot Interchange

The Receive Time-slot Interchange (RTSI) blocks **88**, **90** and **92** re-arrange the constituent STS-1/STM-0 streams of an

STS-48/STM-16 stream in a software configurable order. The RTSI blocks **88**, **90** and **92** also support multi-casting where a STS-1/STM-0 stream from one of the three receive LVDS links is placed on two or more outgoing time-slots. The Receive

5 Working Time-slot Interchange (RWTI) block **88** performs time-slot re-arrangement for data sourced from the working receive LVDS links (RPWRK[4:1]/RNWRK[4:1]). The Received Protection Time-slot Interchange (RPTI) block **90** services the protection receive LVDS links (RPPROT[4:1]/RNPROT[4:1]) while
10 the Receive Auxiliary Time-slot Interchange (RATI) block **92** services the auxiliary receive LVDS links (RPAUX[4:1]/RNAUX[4:1]).

Outgoing TeleCombus PRBS Generator

15 The Outgoing TeleCombus PRBS Generator (OTPG) block **94** optionally inserts a PRBS pattern on a per STS-1/STM-0 onto the Outgoing TeleCombus stream. A total of four OTPG sub-blocks (OTPG #1 to #4) are instantiated in the TBS device. Each OTPG sub-block has the capacity to source PRBS data of
20 an STS-12/STM-4 stream. A set of four OTPG sub-blocks may be connected in tandem to service an STS-48c/STM-16-16c stream. The OTPG block **94** is functionally identical to the generator section of the ITPP block **30**.

LVDS Overview

The LVDS family of cells allow the implementation of 777.6 Mb/s LVDS links. A reference clock of 77.76 MHz is required. Four 777.6 Mb/s LVDS form a set of high-speed serial data links for passing an STS-48 aggregate data stream.

The transmitter drives a differential signal through a pair of 50 Ω characteristic interconnects, such as board traces, backplane traces, or short lengths of cable. The receiver presents a 100 Ω differential termination impedance to terminate the lines. Included in the standard is sufficient common-mode range for the receiver to accommodate as much as 925mV of common-mode ground difference.

Complete SERDES transceiver functionality is provided. Ten-bit parallel data is sampled by the line rate divided-by-10 clock (77.76MHz SYSCLK) and then serialized at the line rate on the LVDS output pins by a 777.6MHz clock synthesized from SYSCLK. Serial line rate LVDS data is sampled and deserialized to 10-bit parallel data. Parallel output transfers are synchronized to a gated line rate divided-by-10 clock. The 10-bit data is passed to an 8B/10B decoding block. The gating duty cycle is adjusted such that the throughput of the parallel interface equals the receive input

data rate (Line Rate +/- 100ppm). It is expected that the clock source of the transmitter and the receiver the same to ensure that the data throughput at both ends of the link are identical.

5

Data must contain sufficient transition density to allow reliable operation of the data recovery units. 8B/10B block coding and decoding is provided by the T8TE and R8TD blocks.

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At the system level, reliable operation will be obtained if proper signal integrity is maintained through the signal path and the receiver requirements are respected. Namely, a worst case eye opening of 0.7UI and 100mV differential amplitude is needed. These conditions should be achievable with a system architecture consisting of board traces, two sets of backplane connectors and up to 1 m of backplane interconnects. This assumes proper design of 100 Ω differential lines and minimization of discontinuities in the signal path. Due to power constraints, the output differential amplitude is approximately 350mV.

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The LVDS system is comprised of the LVDS Receivers (RXLV) **64**, **66** and **68**, LVDS Transmitter (TXLV) **54**, **56** and **58**, Transmitter reference (TXREF) **60**, data recovery units (DRU)

70, 72 and 74, parallel to serial converters (PISO) 48, 50 and 52 and Clock Synthesis Unit (CSU) 62.

Microprocessor Interface

5 The Microprocessor Interface block 98 provides normal and test mode registers, and logic required to connect to the microprocessor interface. The normal mode registers are required for normal operation, and test mode registers are used to enhance testability of the TBS.

Frame Alignment in a Multi-device Environment

10 The RJ0FP frame pulse is used to synchronize a set of devices that are inter-connected via LVDS links. It is provided concurrently to all the devices in the system once every 125 μ s, or multiples thereof. Characters retrieved from the receive LVDS links are written into a FIFO buffer 110. When the J0 character is received, it is written into a fixed location in the FIFO buffer 110. Subsequent characters are written into the locations following the foregoing fixed location. At each device in the system, a software configurable counter is used to mark the point, relative to RJ0FP, where all its receive LVDS links are expected to have delivered their J0 character. As directed by the delay counter, the device will then read the fixed location where the J0 character is stored, thus synchronizing all the

receive LVDS links. Differential delays between and clock instabilities of LVDS links are absorbed by the FIFOs.

This foregoing embodiment provides a new use of 8B/10B control characters to label SONET/SDH transport frame, high-order path frame, and low-order path frame boundaries. The types of bytes that are encoded in 8B/10B control characters are configurable to suit different classes of SONET/SDH equipment (multiplex section terminators, high-order path terminators and low-order path terminators).

Furthermore, the use of 8B/10B encoding on de-scrambled SONET/SDH data streams in order ensures data transitions on the serial links and preserves DC balance.

The standard 8B/10B control character set is extended by treating the positive and negative running disparity codes of those control characters with an even number of ones and zeros as two separate control characters. This implementation doubles the number of control characters available for encoding of events. Further, DC balance is retained without having to alternately send positive and negative running disparity codes.

The use of FIFOs and a universal frame pulse with software programmable delay advantageously allow the transfer

of a single SONET OC-N / SDH STM-M signal over multiple links.

This embodiment also provides a transparent in-band error reporting facility where errors detected at the SONET/SDH receiver can be transferred to the transmitter to construct remote error and defect indication codes.

The occurrence of line code violations of 8B/10B characters can be used to monitor error performance of the serial links.

This embodiment further provides for PRBS pattern insertion and monitoring thereby allowing datapath verification prior to injection of the actual payload.

ALTERNATE EMBODIMENTS

An alternate embodiment utilizes out-of-band signaling to mark frame boundaries, status information and alarm events. SONET/SDH data bytes are carried in byte serial format on a set of four 8-bit buses (ID[4:1][7:0] and OD[4:1][7:0]). Transport frame boundaries and payload frame boundaries are marked by the IJ0J1[4:1] and OJ0J1[4:1] signals. SONET/SDH transport overhead bytes are distinguished from high-order path payload bytes by the IPL[4:1] and OPL[4:1] signals. Low order path payload

boundaries are indicated by the ITV5[4:1] and OTV5[4:1] signals. Low-order path transport overhead bytes are distinguished from low-order path payload bytes by the ITPL[4:1] and OTPL[4:1] signals. Status and alarms are
5 carried by the signals IPAIS[4:1], OPAIS[4:1], ITAIS[4:1] and OTAIS[4:1].

The above-described embodiments should be regarded as illustrative rather than restrictive, and it should be
10 appreciated that variations may be made other than those discussed, by workers of ordinary skill in the art without departing from the scope of the present invention.